

2005 Viterbi Conference

Applications of the Viterbi Algorithm in Data Storage Technology

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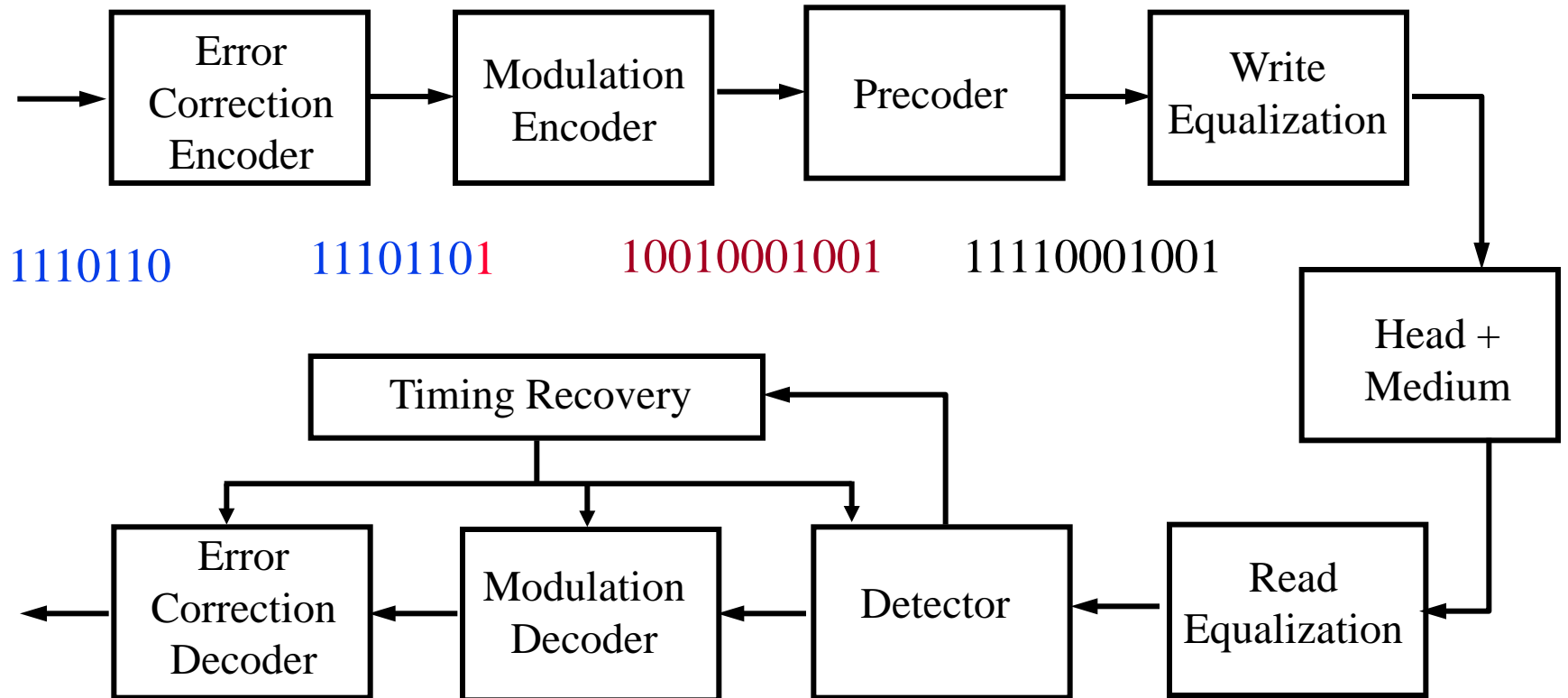
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Outline

- Data storage trends
- Recording channel technology
 - PRML
 - Coded PRML
 - Turbo equalization
- Channel capacity
- Concluding remarks

Digital Recording Channel



Magnetic Recording Process

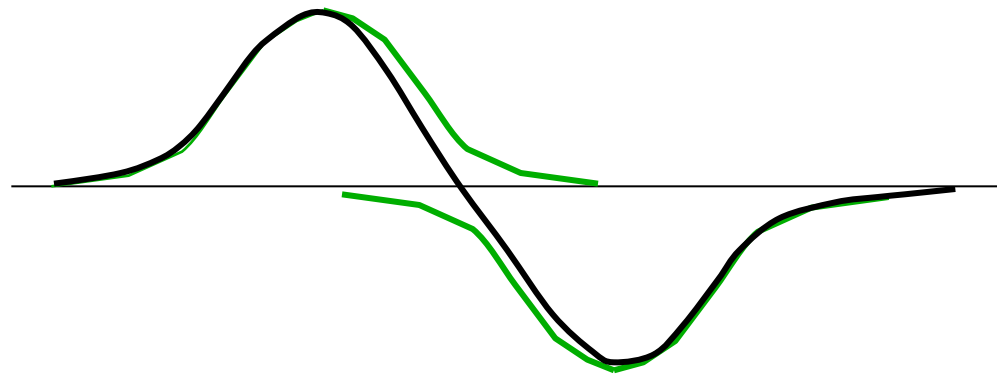
Input signal



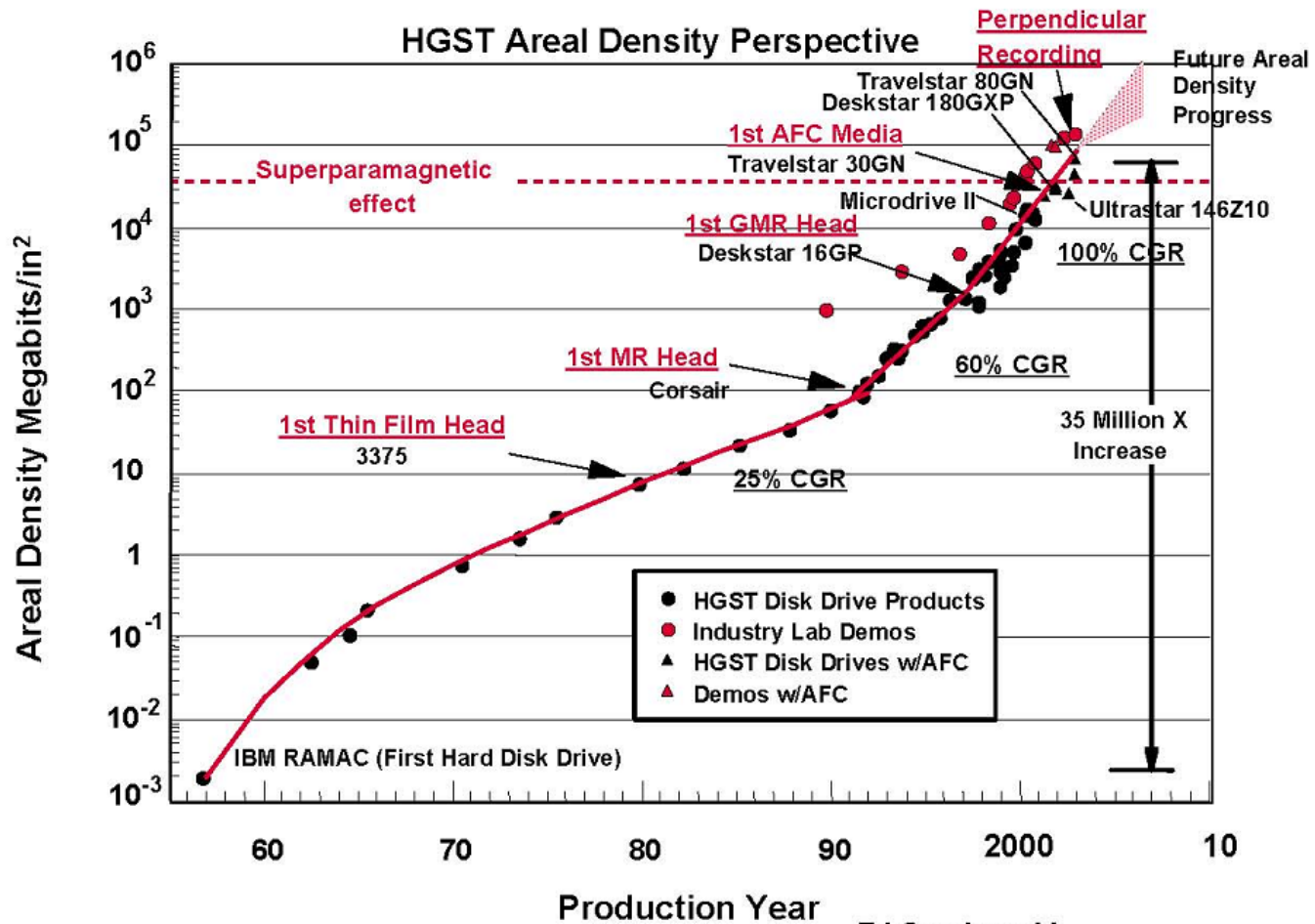
Magnetized
Medium



Readback
Signal



Areal Density Progress

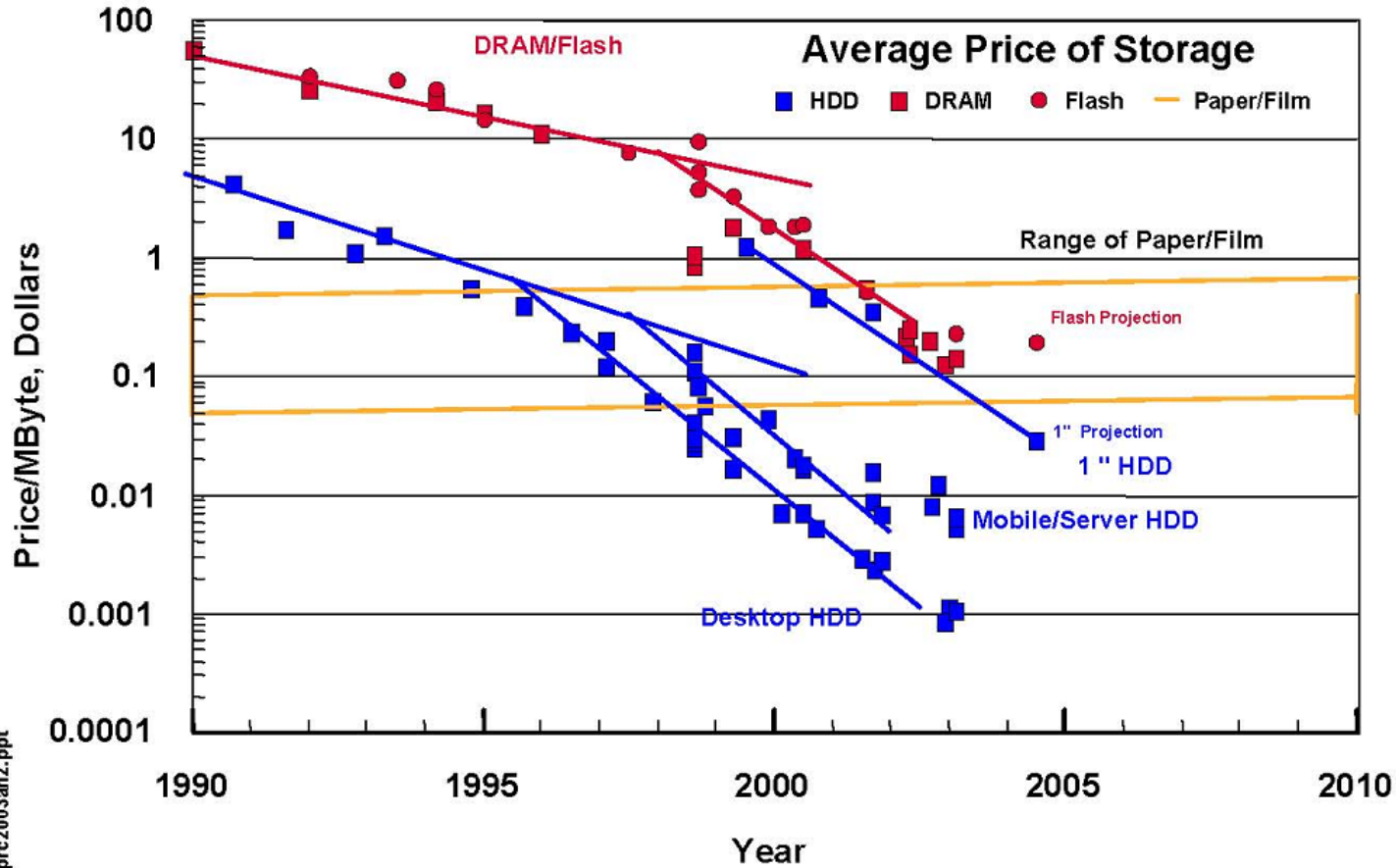


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Average Price of Storage

HITACHI
Inspire the Next



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A Disk Drive (and VA) in Every Pocket

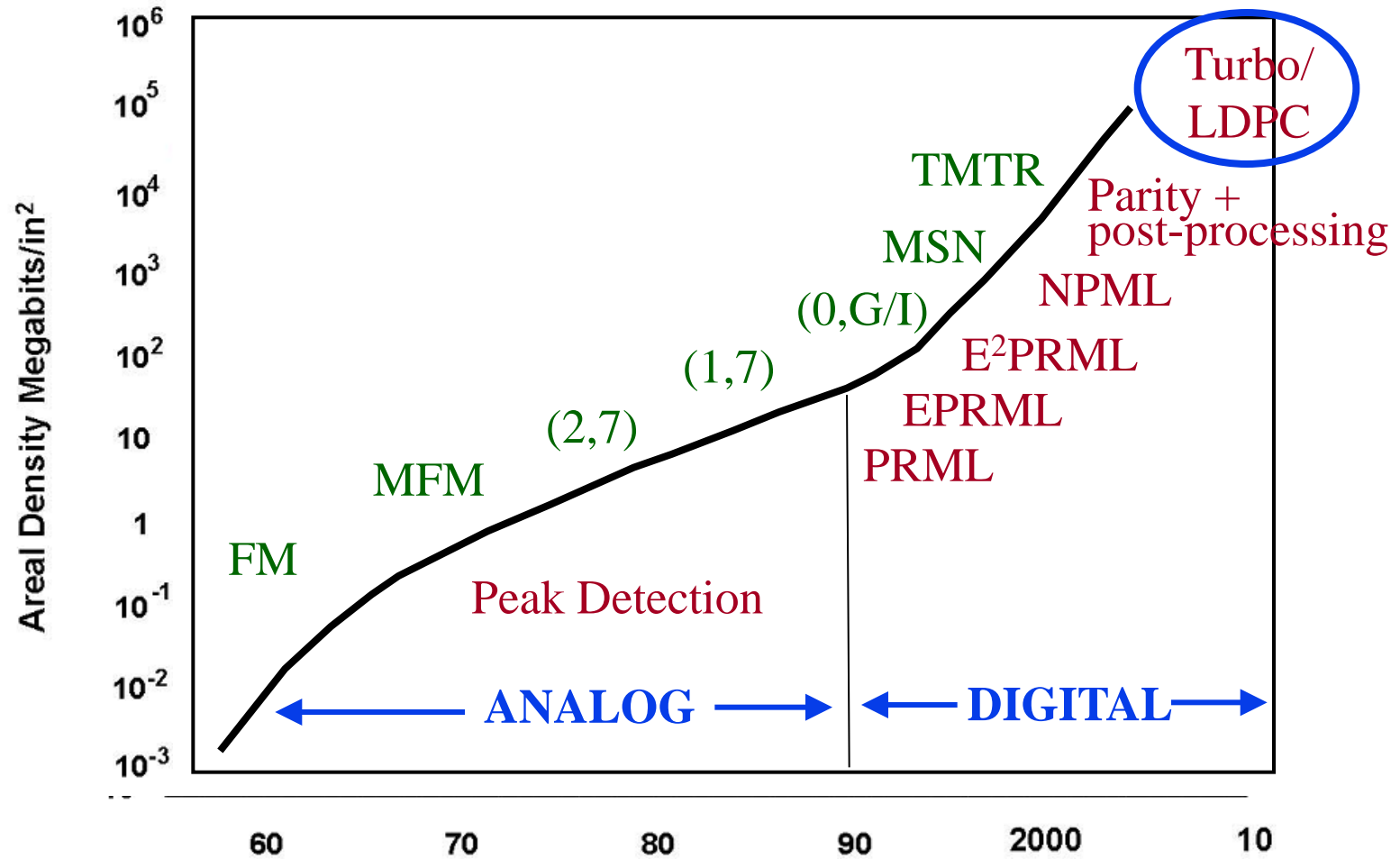


10,000 songs
with album covers



Toshiba 1.8" drive
40.0 Gigabytes
(80GB on the way!)

Signal Processing and Coding Innovation



Key References and Their Impact...

- [1] A.J. Viterbi, “Error Bounds for Convolutional Codes and an Asymptotically Optimum Decoding Algorithm,” *IEEE Transactions on Information Theory*, vol. IT-13, no. 2, pp. 260-269, April 1967.
- [2] A.J. Viterbi, “Convolutional Codes and Their Performance in Communication Systems,” *IEEE Transactions on Communications Technology*, vol. COM-19, no. 5, pp. 751-772, October 1971.
- [3] A.J. Viterbi and J. K. Omura, *Principles of Digital Communication and Coding*. New York, NY: McGraw-Hill, Inc., 1979, Ch. 4.9, pp. 272-284.
- [4] A.J. Viterbi, “An Intuitive Justification and a Simplified Implementation of the MAP Decoder for Convolutional Codes,” *IEEE Journal on Selected Areas in Communications*, vol. 16, no. 2, pp. 260-264, February 1998.

PRML ...

- [1] “Error Bounds for Convolutional Codes and an Asymptotically Optimum Decoding Algorithm”
- **Since the introduction of PRML technology in 1990, the VA has been the standard detection method in disk drives.**

Coded PRML ...

[2] “Convolutional Codes and Their Performance in Communication Systems”

[3] *Principles of Digital Communication and Coding*

- **Since the mid-1990’s, error event characterization of partial-response channels has been used to bound performance and to design constrained modulation codes that detect and/or forbid dominant error events.**

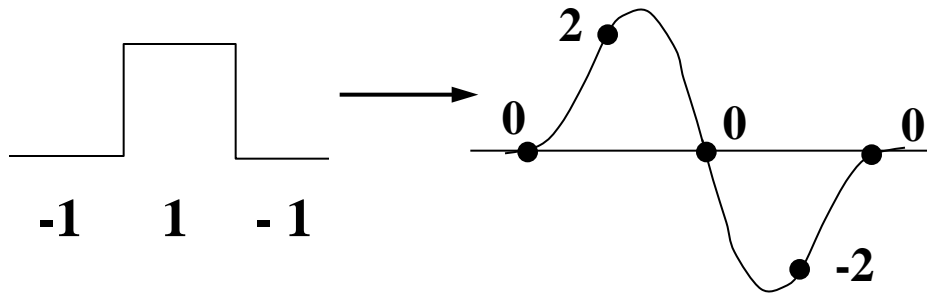
Turbo Equalization and Channel Capacity

[4] “An Intuitive Justification and a Simplified Implementation of the MAP Decoder for Convolutional Codes”

- **“Turbo-equalized” recording channels (proposed) use a modified “dual-max” algorithm for detection and a difference-metric LDPC decoder.**
- **Sharp estimates of the recording channel capacity are calculated using a “generalized VA.”**

What is PRML?

- “PR” = Partial Response [Class-4] Equalization



$$y_n = x_n - x_{n-2}$$

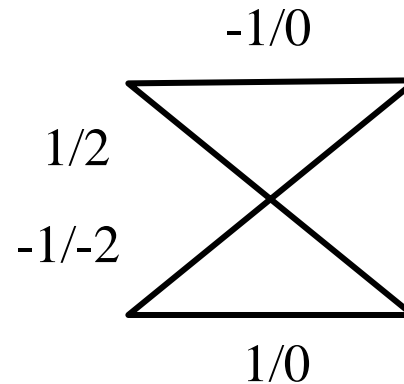
$$h(D) = (1 - D^2) \\ = (1 - D)(1 + D)$$

- “ML” = Maximum Likelihood Sequence Detection (VA)

“Dicode” trellis for even/odd interleaves

$$y_n = x_n - x_{n-1}$$

$$h(D) = 1 - D$$



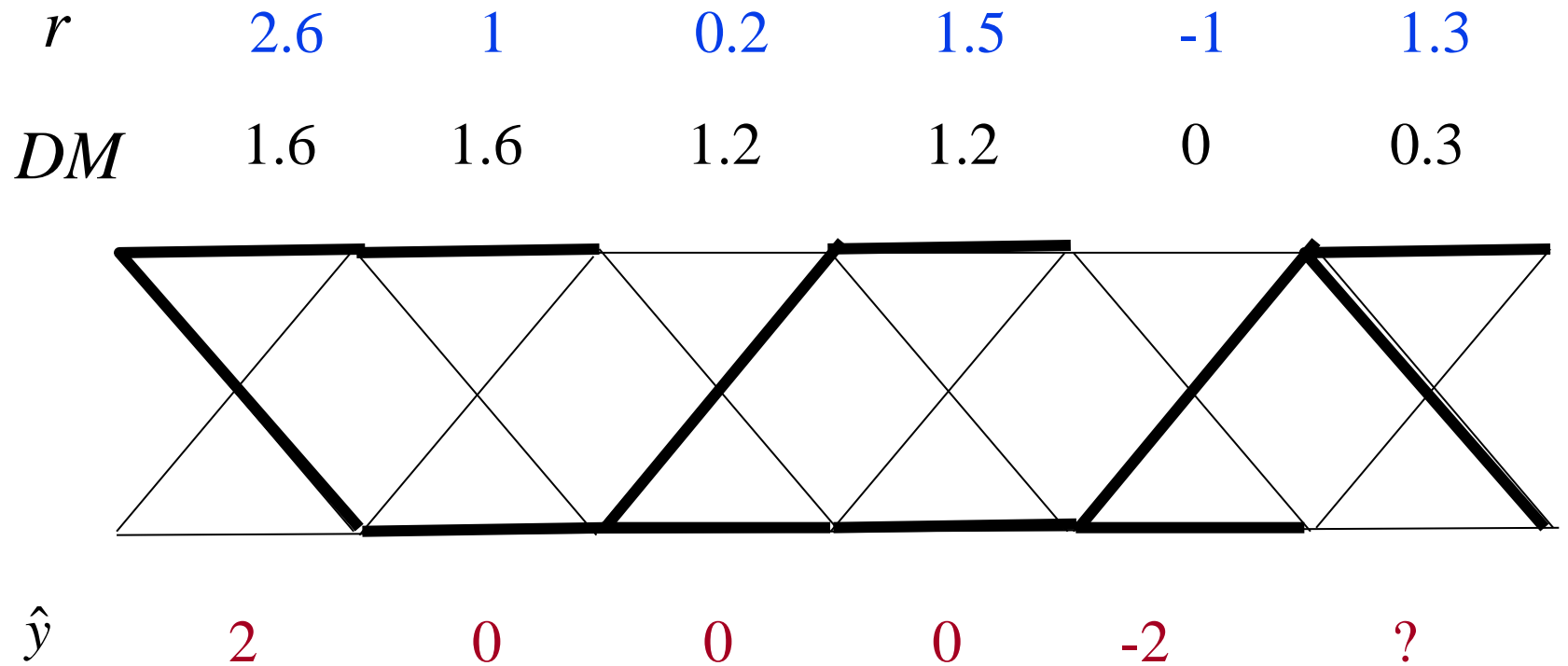
*The acronym “PRML” was coined by Andre Milewski, of IBM LaGaude.

Difference Metric VA for Dicode

$$DM_n = \begin{cases} r_n - 1 & \text{if } DM_{n-2} - r_n \leq -1 & \Rightarrow \text{↘} \\ DM_{n-2} & \text{if } -1 < DM_{n-2} - r_n < 1 & \Rightarrow \text{—} \\ r_n + 1 & \text{if } 1 \leq DM_{n-2} - r_n & \Rightarrow \text{↗} \end{cases}$$

- Used in first commercial disk drive with PRML:
IBM 681 (1990)

Difference Metric VA for Dicode



Beyond PRML

- Extended PRML - “E^NPRML”

$$h(D) = (1 - D)(1 + D)^{N+1}, N \geq 1$$

- Viterbi detector has 2^{N+2} states.
 - EPR4 and E²PR4 have been widely used in commercial drives.
- Noise-predictive PRML (a.k.a. Generalized PRML)

$$h(D) = (1 - D^2)(1 + p_1 D + p_2 D^2)$$



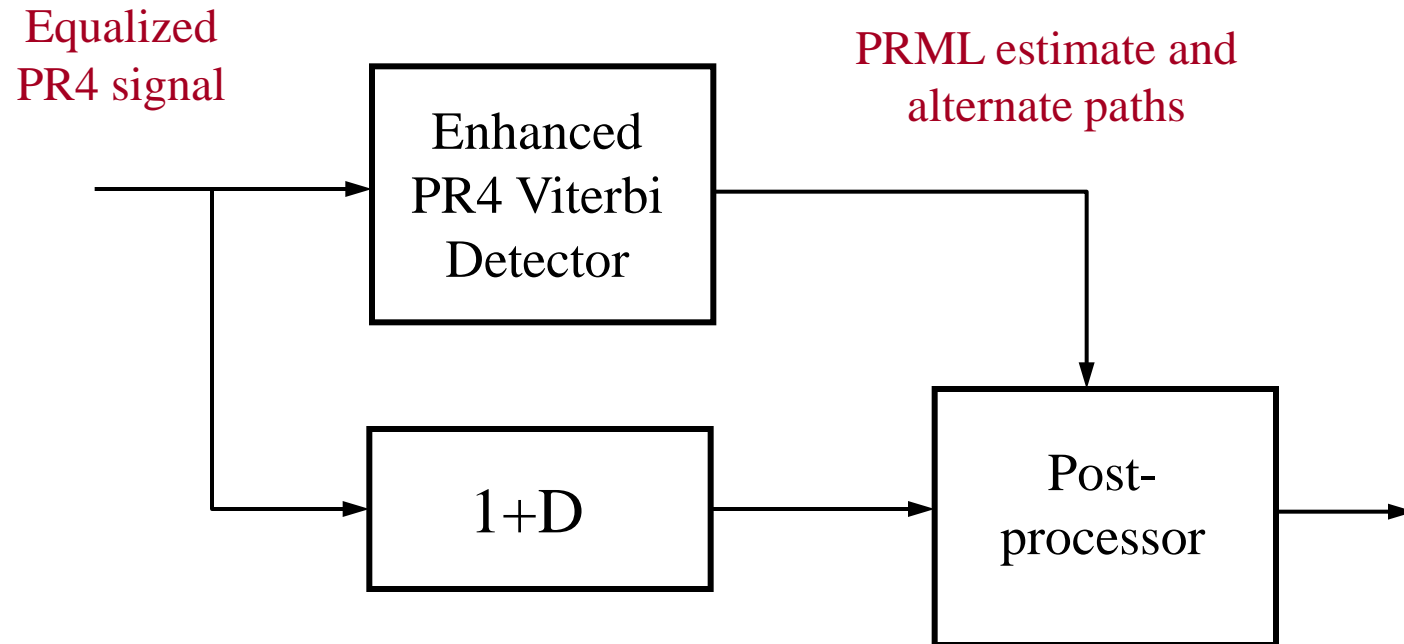
PR4



Noise-whitening filter

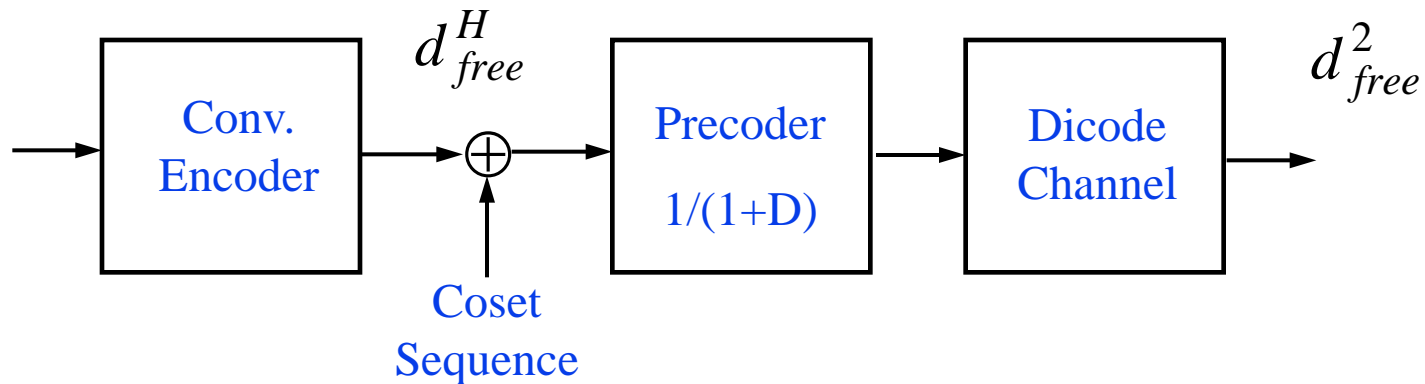
Post-Processor EPRML Detector

- “Turbo-PRML” (1993)



Trellis-coded PRML

- Convolutional code with channel precoder
- Combined convolutional code and channel trellis detector



$$d_{free}^2 \geq \begin{cases} d_{free}^H & \text{if } d_{free}^H \text{ is even} \\ d_{free}^H + 1 & \text{if } d_{free}^H \text{ is odd} \end{cases}$$

Distance-Enhancing Constrained Codes

- Characterize PR channel error-events using error-state diagram analysis. (See [2], [3].)
- Determine modulation constraints that reduce and/or forbid dominant error events, and design code.
- Incorporate channel and code constraints into detector trellis, or use reduced-state trellis and a post-processor.

Error Event Analysis – E²PR4

- E²PR4: $h(D) = (1 - D)(1 + D)^3$

$$d_{free}^2 = 6$$

- Input “error” events: $e(D) = x_1(D) - x_2(D)$

$d^2(e)$	$e = x_1 - x_2$
6	+ - +
8	+ - + 0 0 + - + + - + - (+ -) + - + - (+ -) +

Distance-Enhancing Codes

- Matched-Spectral-Null (MSN) codes

- DC-null and order-K Nyquist null on E²PR4:

$$d_{free}^2 \geq 2(K + 3)$$

- Maximum-Transition-Run MTR(*j,k*) codes

- Limit number of consecutive 1's to *j* (*k*) on even (odd) phase

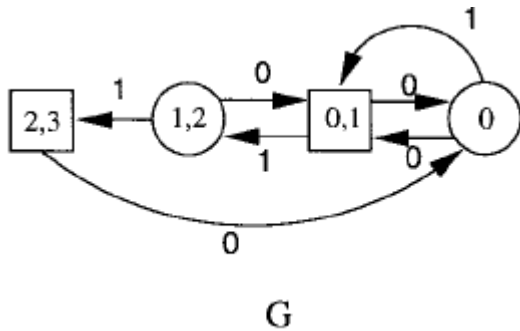
- For E²PR4, the MTR(2,3) constraint yields:

$$d_{free}^2 = 10$$

- Parity-check codes

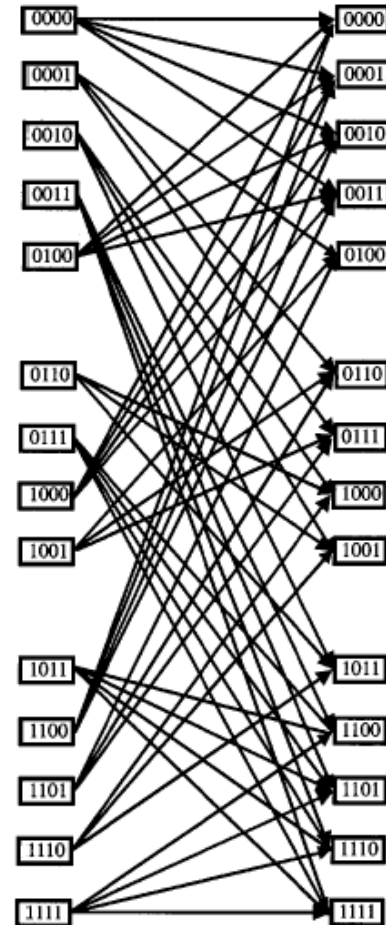
- Detect variety of error events

Combined Code-Channel Trellis



MTR(2,3) constraint graph
(NRZI format)

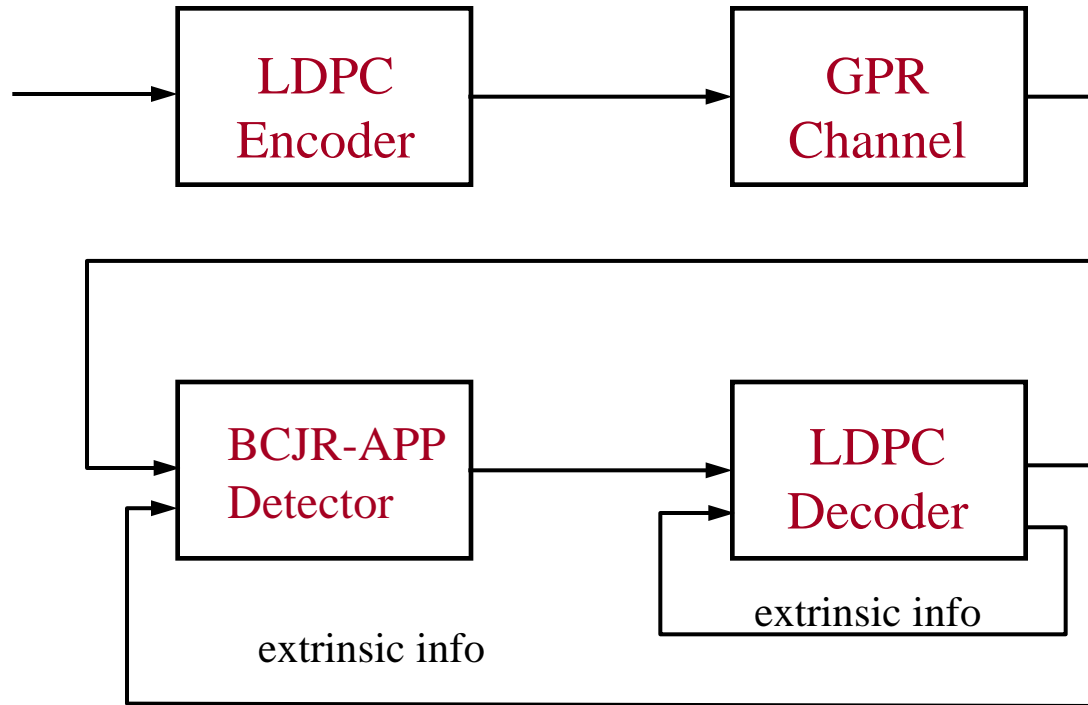
Combined MTR(2,3)
and E²PR4 trellis
(NRZ format)



State-of-the-Art Channel

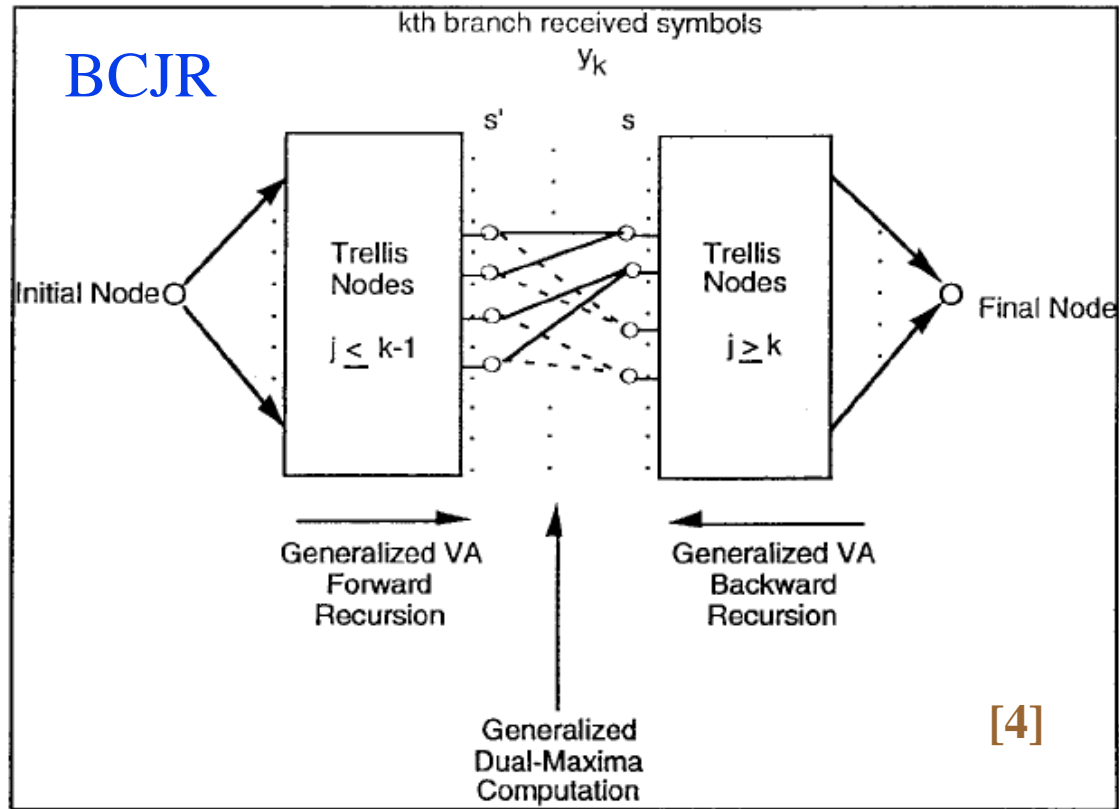
- Rate-96/104 dual-parity code with MTR(3,3) constraints
 - Eliminates all error events of type: $+ - + -$, $+ - + - +$, $+ - + - -$
 - Eliminates half of events of type: $+ - +$
 - Detects error events of type: $+$, $+ -$, $+ - +$, and $+00+$
- 16-state NPML detector with dual-parity post-processing
 - Gain of 0.75dB over rate-48/49, no parity, at $P_e(\text{sector})=10^{-6}$

Turbo Equalization



- Length-4376 LDPC code
- Gain ~4 dB over uncoded NPML at $P_e(\text{symbol})=10^{-5}$
- Gap to capacity ~1.5dB

Simplified BCJR: Dual-Max Detector




$$L_n = \max_{s', s: x=1} \{A_{n-1}(s') + B_n(s) + c_n(s', s)\} - \max_{s', s: x=-1} \{A_{n-1}(s') + B_n(s) + c_n(s', s)\}$$

Capacity of Magnetic Recording Channels

- Binary input, linear ISI, additive, i.i.d. Gaussian noise

$$y[i] = \sum_{k=0}^{n-1} h[k] x[i-k] + n[i]$$

- Capacity C

$$\begin{aligned} C &= \max_{P(X)} I(X;Y) \\ &= \max_{P(X)} H(Y) - H(Y|X) \\ &= \max_{P(X)} H(Y) - \frac{1}{2} \log(\pi e N_0) \end{aligned}$$


For a given $P(X)$, we want to compute $H(Y)$

Computing Entropy Rates

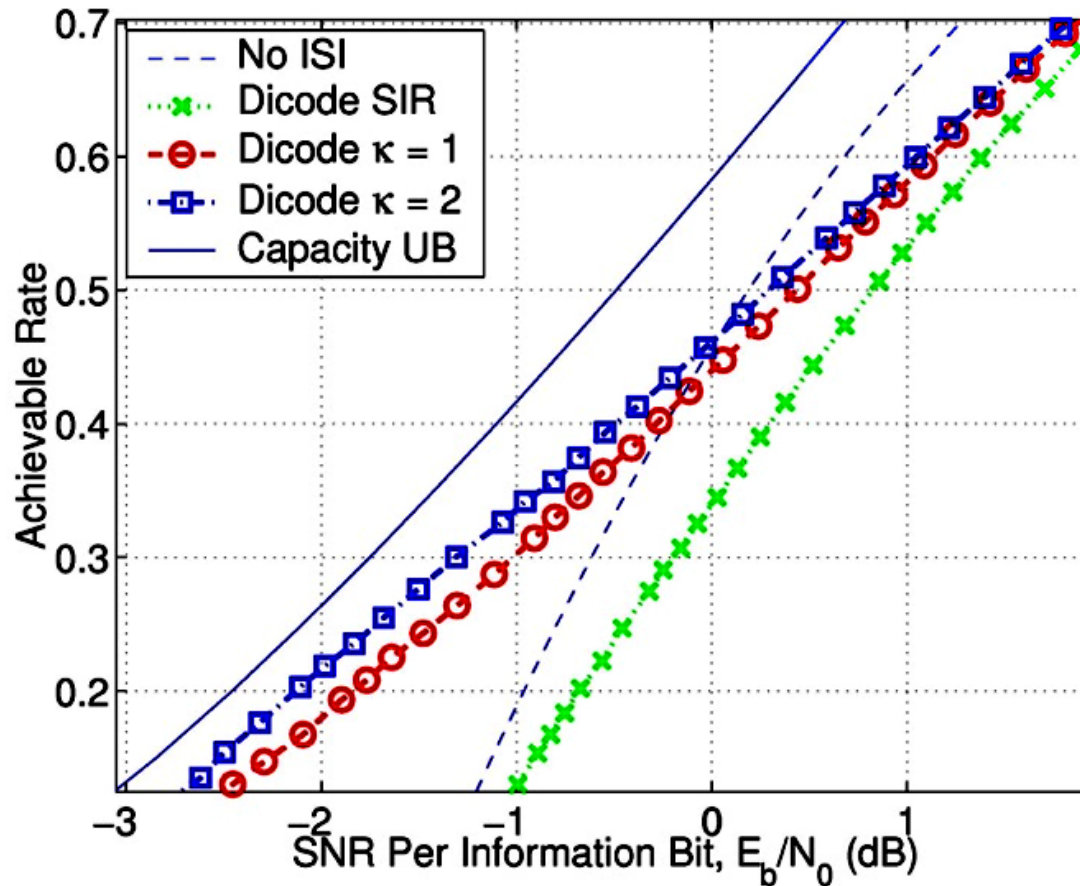
- Shannon-McMillan-Breimann theorem implies

$$-\frac{1}{n} \log p(y_1^n) \xrightarrow{a.s.} H(Y)$$

as $n \rightarrow \infty$, where y_1^n is a single long sample realization of the channel output process.

- The probability $p(y_1^n)$ can be computed using the forward recursion of the BCJR - APP algorithm.
- In the log domain, this forward recursion can be interpreted as a “generalized Viterbi algorithm.” (See [4].)

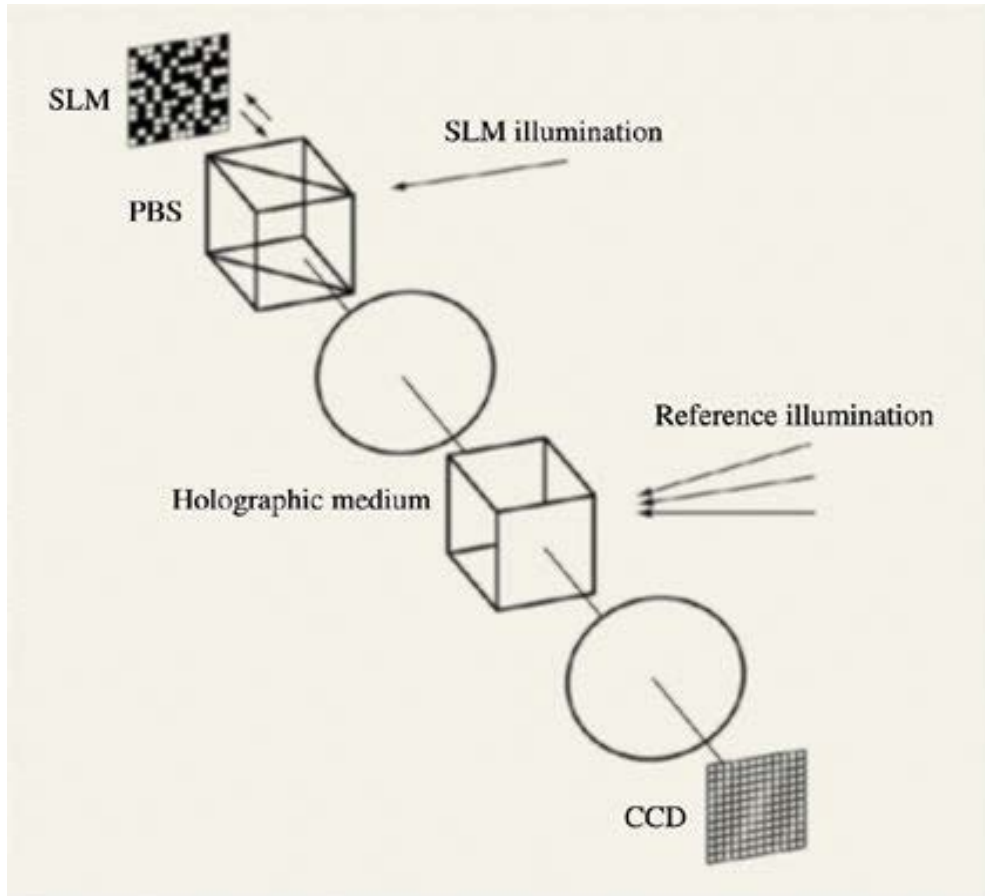
Capacity Bounds for Dicode $h(D)=1-D$



Concluding Remarks

- The Viterbi Algorithm and related ML performance evaluation techniques have been vital to the advancement of data storage technology – magnetic and optical - since 1990.
- The “Viterbi architecture” for APP computation has influenced the development and evaluation of capacity-approaching coding schemes for digital recording applications.
- Future storage technologies offer interesting challenges in detection and decoding...

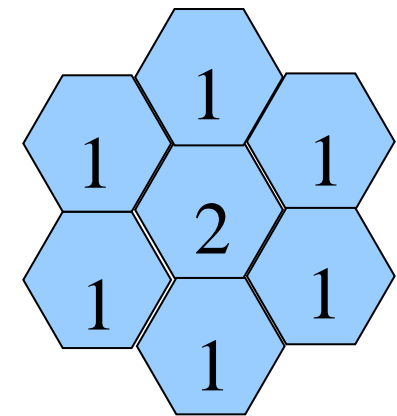
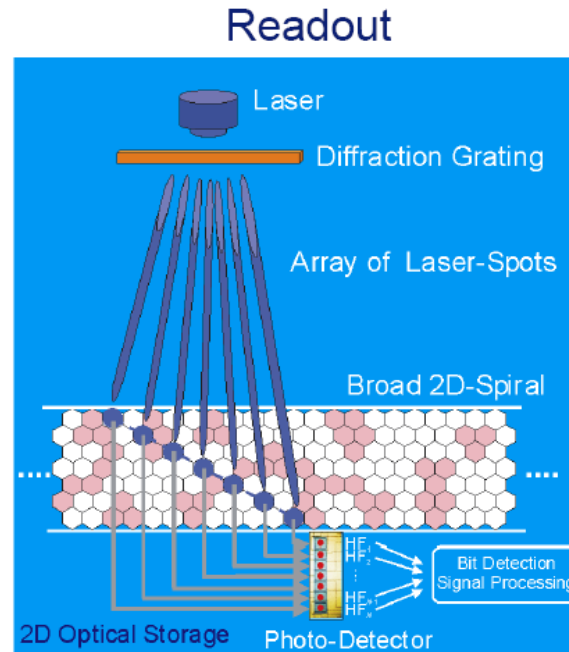
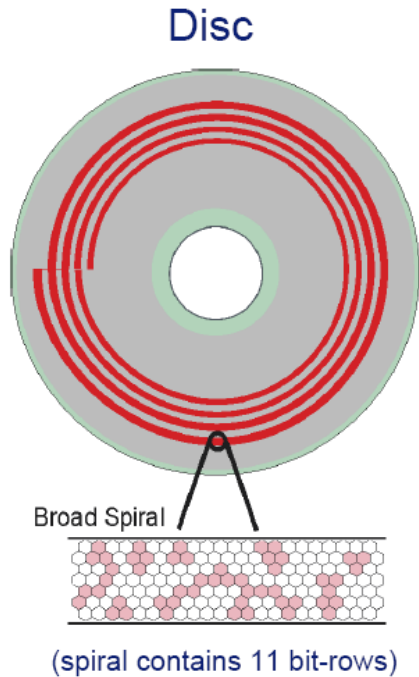
Holographic Recording



2-D Intersymbol Interference

$$h = \begin{bmatrix} 0.05 & 0.21 & 0.05 \\ 0.20 & 0.91 & 0.19 \\ 0.01 & 0.10 & 0.02 \end{bmatrix}$$

Two-Dimensional Optical Storage (TwoDOS)

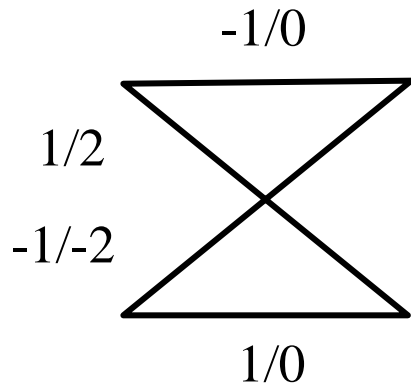


2-D Impulse response

Courtesy of Wim Coene, Philips Research

And, finally...

- **Congratulations – and many thanks – Andy!!**
on the occasion of your milestone birthday, and for your many landmark contributions to science, technology, and engineering education.



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